A Robust FEM-BEM Solver for Time-Harmonic Eddy ² Current Problems ³

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Summary. This paper is devoted to the construction and analysis of robust solution techniques for time-harmonic eddy current problems in unbounded domains. We discretize the time-harmonic eddy current equation by means of a symmetrically coupled finite and boundary element method, taking care of the different physical behavior in conducting and nonconducting subdomains, respectively. We construct and analyse a block-diagonal preconditioner for the system of coupled finite and boundary element equations that is robust with respect to the space discretization parameter as well as all involved "bad" parameters like the frequency, the conductivity and the reluctivity. Block-diagonal preconditioners can be used for accelerating iterative solution methods such like the Minimal Residual Method.

1 Introduction

In many practical applications, the excitation is time-harmonic. Switching from the 19 time domain to the frequency domain allows us to replace expensive time-integration 20 procedures by the solution of a system of partial differential equations for the am- 21 plitudes belonging to the sine- and to the cosine-excitation. Following this strat- 22 egy, [7, 13] and [4, 5] applied harmonic and multiharmonic approaches to parabolic 23 initial-boundary value problems and the eddy current problem, respectively. Indeed, 24 in [13], a preconditioned MinRes solver for the solution of the eddy current problem 25 in bounded domains was constructed that is robust with respect to both the discretiza- 26 tion parameter h and the frequency ω . The key point of this parameter-robust solver 27 is the construction of a block-diagonal preconditioner, where standard H(curl) FEM 28 magneto-static problems have to be solved or preconditioned. The aim of this con- 29 tribution is to generalize these ideas to the case of unbounded domains in terms of 30 a coupled Finite Element (FEM) - Boundary Element (BEM) Method. In this case 31 we are also able to construct a block-diagonal preconditioner, where now standard 32 coupled FEM-BEM H(curl) problems, as arising in the magneto-static case, have 33 to be solved or preconditioned. We mention, that this preconditioning technique fits 34 into the framework of operator preconditioning, see, e.g. [1, 11, 16, 19]. 35

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The paper is now organized as follows. We introduce the frequency domain ³⁶ equations in Sect. 2. In the same section, we provide the symmetrically coupled ³⁷ FEM-BEM discretization of these equations. In Sect. 3, we construct and analyse our ³⁸ parameter-robust block-diagonal preconditioner used in a MinRes setting for solving ³⁹ the resulting system of linear algebraic equations. Finally, we discuss the practical ⁴⁰ realization of our preconditioner.

2 Frequency Domain FEM-BEM

As a model problem, we consider the following eddy current problem:

$$\begin{cases} \sigma \frac{\partial \mathbf{u}}{\partial t} + \mathbf{curl} \left(v_1 \, \mathbf{curl} \, \mathbf{u} \right) = \mathbf{f} & \text{in } \Omega_1 \times (0, T), \\ \mathbf{curl} \left(\mathbf{curl} \, \mathbf{u} \right) = \mathbf{0} & \text{in } \Omega_2 \times (0, T), \\ \text{div } \mathbf{u} = 0 & \text{in } \Omega_2 \times (0, T), \\ \mathbf{u} = \mathcal{O}(|\mathbf{x}|^{-1}) & \text{for } |\mathbf{x}| \to \infty, \\ \mathbf{curl} \, \mathbf{u} = \mathcal{O}(|\mathbf{x}|^{-1}) & \text{for } |\mathbf{x}| \to \infty, \\ \mathbf{u} = \mathbf{u}_0 & \text{on } \Omega_1 \times \{\mathbf{0}\}, \\ \mathbf{u}_1 \times \mathbf{n} = \mathbf{u}_2 \times \mathbf{n} & \text{on } \Gamma \times (0, T), \\ v_1 \mathbf{curl} \, \mathbf{u}_1 \times \mathbf{n} = \mathbf{curl} \, \mathbf{u}_2 \times \mathbf{n} \text{ on } \Gamma \times (0, T), \end{cases}$$
(1)

where the computational domain $\Omega = \mathbb{R}^3$ is split into the two non-overlapping subdomains Ω_1 and Ω_2 . The conducting subdomain Ω_1 is assumed to be a simply 45 connected Lipschitz polyhedron, whereas the non-conducting subdomain Ω_2 is the 46 complement of Ω_1 in \mathbb{R}^3 , i.e $\mathbb{R}^3 \setminus \overline{\Omega_1}$. Furthermore, we denote by Γ the interface between the two subdomains, i.e. $\Gamma = \overline{\Omega_1} \cap \overline{\Omega_2}$. The exterior unit normal vector of Ω_1 48 on Γ is denoted by **n**, i.e. **n** points from Ω_1 to Ω_2 . The reluctivity v_1 is supposed to be independent of $|\mathbf{curlu}|$, i.e. we assume the eddy current problem (1) to be linear. The conductivity σ is zero in Ω_2 , and piecewise constant and uniformly positive in Ω_1 .

We assume, that the source **f** is given by a time-harmonic excitation with the ⁵² frequency $\omega > 0$ and amplitudes **f**^c and **f**^s in the conducting domain Ω_1 . Therefore, ⁵³ the solution **u** is time-harmonic as well, with the same base frequency ω , i.e. ⁵⁴

$$\mathbf{u}(\mathbf{x},t) = \mathbf{u}^{\mathbf{c}}(\mathbf{x})\cos(\omega t) + \mathbf{u}^{\mathbf{s}}(\mathbf{x})\sin(\omega t).$$
(2)

In fact, (2) is the real reformulation of a complex time-harmonic approach $\mathbf{u}(\mathbf{x},t) = 55$ $\mathbf{\hat{u}}(\mathbf{x})e^{i\omega t}$ with the complex-valued amplitude $\mathbf{\hat{u}} = \mathbf{u}^{c} - i\mathbf{u}^{s}$. Using the time-harmonic 56 representation (2) of the solution, we can state the eddy current problem (1) in the 57 frequency domain as follows: 58

Find
$$\mathbf{u} = (\mathbf{u}^{\mathbf{c}}, \mathbf{u}^{\mathbf{s}}):$$

$$\begin{cases}
\omega \sigma \mathbf{u}^{\mathbf{s}} + \operatorname{curl}(v_{1} \operatorname{curl} \mathbf{u}^{\mathbf{c}}) = \mathbf{f}^{\mathbf{c}} \text{ in } \Omega_{1}, \\
\operatorname{curl} \operatorname{curl} \mathbf{u}^{\mathbf{c}} = \mathbf{0} \text{ in } \Omega_{2}, \\
-\omega \sigma \mathbf{u}^{\mathbf{c}} + \operatorname{curl}(v_{1} \operatorname{curl} \mathbf{u}^{\mathbf{s}}) = \mathbf{f}^{\mathbf{s}} \text{ in } \Omega_{1}, \\
\operatorname{curl} \operatorname{curl} \mathbf{u}^{\mathbf{s}} = \mathbf{0} \text{ in } \Omega_{2},
\end{cases}$$
(3)

with the corresponding decay and interface conditions from (1).

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Remark 1. In practice, the reluctivity v_1 depends on the inductivity $|\mathbf{curl u}|$ in a non- ⁶⁰ linear way in ferromagnetic materials. Having in mind applications to problems with ⁶¹ nonlinear reluctivity, we prefer to use the real reformulation (3) instead of a complex ⁶² approach. For overcoming the nonlinearity the preferable way is to apply Newton's ⁶³ method due to its fast convergence. It turns out, that Newton's method cannot be applied to the nonlinear complex-valued system (see [4]), but it can be applied to the ⁶⁵ reformulated real-valued system. Anyhow, the analysis of the linear problem also ⁶⁶ helps to construct efficient solvers for the nonlinear problem. ⁶⁷

Deriving the variational formulation and integrating by parts once more in the exterior domain yields: Find $(\mathbf{u}^{c}, \mathbf{u}^{s}) \in \mathbf{H}(\mathbf{curl}, \Omega_{1})^{2}$ such that

$$\begin{cases} \omega(\sigma \mathbf{u}^{\mathbf{s}}, \mathbf{v}^{\mathbf{c}})_{L_{2}(\Omega_{1})} + (\nu_{1} \mathbf{curl} \mathbf{u}^{\mathbf{c}}, \mathbf{curl} \mathbf{v}^{\mathbf{c}})_{L_{2}(\Omega_{1})} - \langle \gamma_{N} \mathbf{u}^{\mathbf{c}}, \gamma_{D} \mathbf{v}^{\mathbf{c}} \rangle_{\tau} = \langle \mathbf{f}^{\mathbf{c}}, \mathbf{v}^{\mathbf{c}} \rangle, \\ -\omega(\sigma \mathbf{u}^{\mathbf{c}}, \mathbf{v}^{\mathbf{s}})_{L_{2}(\Omega_{1})} + (\nu_{1} \mathbf{curl} \mathbf{u}^{\mathbf{s}}, \mathbf{curl} \mathbf{v}^{\mathbf{s}})_{L_{2}(\Omega_{1})} - \langle \gamma_{N} \mathbf{u}^{\mathbf{s}}, \gamma_{D} \mathbf{v}^{\mathbf{s}} \rangle_{\tau} = \langle \mathbf{f}^{\mathbf{s}}, \mathbf{v}^{\mathbf{s}} \rangle, \end{cases}$$

for all $(\mathbf{v}^{\mathbf{c}}, \mathbf{v}^{\mathbf{s}}) \in \mathbf{H}(\mathbf{curl}, \Omega_1)^2$. Here γ_D and γ_N denote the Dirichlet trace $\gamma_D := \mathbf{n} \times 70$ $(\mathbf{u} \times \mathbf{n})$ and the Neumann trace $\gamma_N := \mathbf{curl} \mathbf{u} \times \mathbf{n}$ on the interface Γ . $\langle \cdot, \cdot \rangle_{\tau}$ denotes the 71 $L_2(\Gamma)$ -based duality product. In order to deal with the expression on the interface 72 Γ , we use the framework of the symmetric FEM-BEM coupling for eddy current 73 problems (see [10]). So, using the boundary integral operators \mathbf{A} , \mathbf{B} , \mathbf{C} and \mathbf{N} , as 74 defined in [10], we end up with the weak formulation of the time-harmonic eddy 75 current problem: Find $(\mathbf{u}^{\mathbf{c}}, \mathbf{u}^{\mathbf{s}}) \in \mathbf{H}(\mathbf{curl}, \Omega_1)^2$ and $(\lambda^{\mathbf{c}}, \lambda^{\mathbf{s}}) \in \mathbf{H}_{\parallel}^{-\frac{1}{2}}(\operatorname{div}_{\Gamma} 0, \Gamma)^2$ such 76 that

$$\begin{cases} \omega(\sigma \mathbf{u}^{\mathbf{s}}, \mathbf{v}^{\mathbf{c}})_{L_{2}(\Omega_{1})} + (v_{1} \operatorname{\mathbf{curl}} \mathbf{u}^{\mathbf{c}}, \operatorname{\mathbf{curl}} \mathbf{v}^{\mathbf{c}})_{L_{2}(\Omega_{1})}, \\ -\langle \mathbf{N}(\gamma_{D}\mathbf{u}^{\mathbf{c}}), \gamma_{D}\mathbf{v}^{\mathbf{c}}\rangle_{\tau} + \langle \mathbf{B}(\lambda^{\mathbf{c}}), \gamma_{D}\mathbf{v}^{\mathbf{c}}\rangle_{\tau} = \langle \mathbf{f}^{\mathbf{c}}, \mathbf{v}^{\mathbf{c}}\rangle, \\ \langle \mu^{\mathbf{c}}, (\mathbf{C} - \mathbf{Id})(\gamma_{D}\mathbf{u}^{\mathbf{c}})\rangle_{\tau} - \langle \mu^{\mathbf{c}}, \mathbf{A}(\lambda^{\mathbf{c}})\rangle_{\tau} = 0, \\ -\omega(\sigma \mathbf{u}^{\mathbf{c}}, \mathbf{v}^{\mathbf{s}})_{L_{2}(\Omega_{1})} + (v_{1} \operatorname{\mathbf{curl}} \mathbf{u}^{\mathbf{s}}, \operatorname{\mathbf{curl}} \mathbf{v}^{\mathbf{s}})_{L_{2}(\Omega_{1})}, \\ -\langle \mathbf{N}(\gamma_{D}\mathbf{u}^{\mathbf{s}}), \gamma_{D}\mathbf{v}^{\mathbf{s}}\rangle_{\tau} + \langle \mathbf{B}(\lambda^{\mathbf{s}}), \gamma_{D}\mathbf{v}^{\mathbf{s}}\rangle_{\tau} = \langle \mathbf{f}^{\mathbf{s}}, \mathbf{v}^{\mathbf{s}}\rangle, \\ \langle \mu^{\mathbf{s}}, (\mathbf{C} - \mathbf{Id})(\gamma_{D}\mathbf{u}^{\mathbf{s}})\rangle_{\tau} - \langle \mu^{\mathbf{s}}, \mathbf{A}(\lambda^{\mathbf{s}})\rangle_{\tau} = 0, \end{cases}$$
(4)

for all $(\mathbf{v}^{\mathbf{c}}, \mathbf{v}^{\mathbf{s}}) \in \mathbf{H}(\mathbf{curl}, \Omega_1)^2$ and $(\mu^{\mathbf{c}}, \mu^{\mathbf{s}}) \in \mathbf{H}_{\parallel}^{-\frac{1}{2}}(\operatorname{div}_{\Gamma}0, \Gamma)^2$. This variational form 78 is the starting point of the discretization in space. Therefore, we use a regular trian-79 gulation \mathcal{T}_h , with mesh size h > 0, of the domain Ω_1 with tetrahedral elements. \mathcal{T}_h 80 induces a mesh \mathcal{H}_h of triangles on the boundary Γ . On these meshes, we consider 81 Nédélec basis functions of order p yielding the conforming finite element subspace 82 $\mathcal{ND}_p(\mathcal{T}_h)$ of $\mathbf{H}(\mathbf{curl}, \Omega_1)$, see [17]. Further, we use the space of divergence free 83 Raviart-Thomas basis functions $\mathcal{RT}_p^0(\mathcal{H}_h) := \{\lambda_h \in \mathcal{RT}_p(\mathcal{H}_h), \operatorname{div}_{\Gamma}\lambda_h = 0\}$ being 84 a conforming finite element subspace of $\mathbf{H}_{\parallel}^{-\frac{1}{2}}(\operatorname{div}_{\Gamma}0, \Gamma)$. Let $\{\varphi_i\}$ denote the basis 85 of $\mathcal{ND}_p(\mathcal{T}_h)$, and let $\{\psi_i\}$ denote the basis of $\mathcal{RT}_p^0(\mathcal{H}_h)$. Then the matrix entries 86 corresponding to the operators in (4) are given by the formulas

$$\begin{split} (\mathbf{K})_{ij} &:= (\nu \operatorname{curl} \varphi_{\mathbf{i}}, \operatorname{curl} \varphi_{\mathbf{j}})_{\mathbf{L}_{2}(\Omega_{1})} - \langle \mathbf{N}(\gamma_{D}\varphi_{\mathbf{i}}), \gamma_{D}\varphi_{\mathbf{j}} \rangle_{\tau}, \\ (\mathbf{M})_{ij} &:= \omega(\sigma\varphi_{\mathbf{i}}, \varphi_{\mathbf{j}})_{\mathbf{L}_{2}(\Omega_{1})}, \\ (\mathbf{A})_{ij} &:= \langle \psi_{\mathbf{i}}, \mathbf{A}(\psi_{\mathbf{j}}) \rangle_{\tau}, \\ (\mathbf{B})_{ij} &:= \langle \psi_{\mathbf{i}}, (\mathbf{C} - \operatorname{Id})(\gamma_{D}\varphi_{\mathbf{j}}) \rangle_{\tau}. \end{split}$$

The entries of the right-hand side vector are given by the formulas $(\mathbf{f}^{\mathbf{c}})_i := 88$ $(\mathbf{f}^{\mathbf{c}}, \varphi_{\mathbf{i}})_{\mathbf{L}_2(\Omega_1)}$ and $(\mathbf{f}^{\mathbf{s}})_i := (\mathbf{f}^{\mathbf{s}}, \varphi_{\mathbf{i}})_{\mathbf{L}_2(\Omega_1)}$. The resulting system $\mathscr{A} \mathbf{x} = \mathbf{f}$ of the coupled 89 finite and boundary element equations has now the following structure: 90

$$\begin{pmatrix} \mathbf{M} & 0 & \mathbf{K} & \mathbf{B}^T \\ 0 & 0 & \mathbf{B} & -\mathbf{A} \\ \mathbf{K} & \mathbf{B}^T & -\mathbf{M} & 0 \\ \mathbf{B} & -\mathbf{A} & 0 & 0 \end{pmatrix} \begin{pmatrix} \mathbf{u}^s \\ \lambda^s \\ \mathbf{u}^c \\ \lambda^c \end{pmatrix} = \begin{pmatrix} \mathbf{f}^c \\ 0 \\ \mathbf{f}^s \\ 0 \end{pmatrix}.$$
 (5)

In fact, the system matrix \mathscr{A} is symmetric and indefinite and obtains a double 91 saddle-point structure. Since \mathscr{A} is symmetric, the system can be solved by a Min- 92 Res method, see, e.g., [18]. Anyhow, the convergence rate of any iterative method 93 deteriorates with respect to the meshsize *h* and the "bad" parameters ω , *v* and σ , 94 if applied to the unpreconditioned system (5). Therefore, preconditioning is a chal-95 lenging topic. 96

3 A Parameter-Robust Preconditioning Technique

In this section, we investigate a preconditioning technique for double saddle-point 98 equations with the block-structure (5). Due to the symmetry and coercivity properties 99 of the underlying operators, the blocks fulfill the following properties: $\mathbf{K} = \mathbf{K}^T \ge 0$, 100 $\mathbf{M} = \mathbf{M}^T > 0$ and $\mathbf{A} = \mathbf{A}^T > 0$.

In [19] a parameter-robust block-diagonal preconditioner for the distributed optimal control of the Stokes equations is constructed. The structural similarities to that preconditioner gives us a hint how to choose the block-diagonal preconditioner in our case. Therefore, we propose the following preconditioner 105

$$\mathscr{C} = \operatorname{diag} \left(\mathscr{I}_{FEM}, \mathscr{I}_{BEM}, \mathscr{I}_{FEM}, \mathscr{I}_{BEM} \right),$$

where the diagonal blocks are given by $\mathscr{I}_{FEM} = \mathbf{M} + \mathbf{K}$ and $\mathscr{I}_{BEM} = \mathbf{A} + \mathbf{B} \mathscr{I}_{FEM}^{-1} \mathbf{B}^T$. 106 Being aware that \mathscr{I}_{FEM} and \mathscr{I}_{BEM} are symmetric and positive definite, we conclude 107 that \mathscr{C} is also symmetric and positive definite. Therefore, \mathscr{C} induces the energy norm 108 $\|\mathbf{u}\|_{\mathscr{C}} = \sqrt{\mathbf{u}^T \mathscr{C} \mathbf{u}}$. Using this special norm, we can apply the Theorem of Babuška-Aziz [3] to the variational problem: 110

Find
$$\mathbf{x} \in \mathbb{R}^N$$
: $\mathbf{w}^T \mathscr{A} \mathbf{x} = \mathbf{w}^T \mathbf{f}, \quad \forall \mathbf{w} \in \mathbb{R}^N.$

The main result is now summarized in the following lemma.

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Lemma 1. The matrix \mathcal{A} satisfies the following norm equivalence inequalities: 112

$$\frac{1}{\sqrt{7}} \|\mathbf{x}\|_{\mathscr{C}} \leq \sup_{\mathbf{w}\neq 0} \frac{\mathbf{w}^T \mathscr{A} \mathbf{x}}{\|\mathbf{w}\|_{\mathscr{C}}} \leq 2 \, \|\mathbf{x}\|_{\mathscr{C}} \quad \forall \mathbf{x} \in \mathbb{R}^N.$$

Proof. Throughout the proof, we use the following notation: $\mathbf{x} = (\mathbf{x_1}, \mathbf{x_2}, \mathbf{x_3}, \mathbf{x_4})^T$ 113 and $\mathbf{y} = (\mathbf{y}_1, \mathbf{y}_2, \mathbf{y}_3, \mathbf{y}_4)^T$. The upper bound follows by reapplication of Cauchy's inequality several time. The expressions corresponding to the Schur complement can 115 be derived in the following way: 116

$$\mathbf{y_1}^T \mathbf{B}^T \mathbf{x_4} = \mathbf{y_1} \mathscr{I}_{FEM}^{1/2} \mathscr{I}_{FEM}^{-1/2} \mathbf{B}^T \mathbf{x_4} \le \|\mathscr{I}_{FEM}^{1/2} \mathbf{y_1}\|_{l_2} \|\mathscr{I}_{FEM}^{-1/2} \mathbf{B}^T \mathbf{x_4}\|_{l_2}.$$

Therefore, we end up with an upper bound with constant 2.

In order to compute the lower bound, we use a linear combination of special test 118 vectors. For the choice $\mathbf{w_1} = (\mathbf{x_1}, \mathbf{x_2}, -\mathbf{x_3}, -\mathbf{x_4})^T$, we obtain 119

$$\mathbf{w_1}^T \mathscr{A} \mathbf{x} = \mathbf{x_1}^T \mathbf{M} \mathbf{x_1} + \mathbf{x_3}^T \mathbf{M} \mathbf{x_3};$$

for $\mathbf{w_2} = (\mathbf{x_3}, -\mathbf{x_4}, \mathbf{x_1}, -\mathbf{x_2})^T$, we get

 $\mathbf{x}_4, \mathbf{x}_1, -\mathbf{x}_2)^T$, we get $\mathbf{w}_2^T \mathscr{A} \mathbf{x} = \mathbf{x}_1^T \mathbf{K} \mathbf{x}_1 + \mathbf{x}_3^T \mathbf{K} \mathbf{x}_3 + \mathbf{x}_2^T \mathbf{A} \mathbf{x}_2 + \mathbf{x}_4^T \mathbf{A} \mathbf{x}_4;$

for
$$\mathbf{w}_3 = ((\mathbf{x_4}^T \mathbf{B} (\mathbf{K} + \mathbf{M})^{-1})^T, \mathbf{0}, (\mathbf{x_2}^T \mathbf{B} (\mathbf{K} + \mathbf{M})^{-1})^T, \mathbf{0})^T$$
, we have

$$\begin{split} \mathbf{w_3}^T \mathscr{A} \mathbf{x} &= \mathbf{x_4}^T \mathbf{B} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{B}^T \mathbf{x_4} + \mathbf{x_2}^T \mathbf{B} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{B}^T \mathbf{x_2} \\ &+ \mathbf{x_4}^T \mathbf{B} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_1} + \mathbf{x_4}^T \mathbf{B} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x_3} \\ &+ \mathbf{x_2}^T \mathbf{B} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x_1} - \mathbf{x_2}^T \mathbf{B} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_3}; \end{split}$$

for
$$\mathbf{w}_4 = (-(\mathbf{x}_3^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1})^T, \mathbf{0}, -(\mathbf{x}_1^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1})^T, \mathbf{0})^T$$
, we get
 $\mathbf{w}_4^T \mathscr{A} \mathbf{x} = -\mathbf{x}_3^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x}_1 - \mathbf{x}_3^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x}_3$

$$\begin{array}{l} {}_{4} \mathcal{P} \,\mathcal{A} \,\mathbf{x} = & -\mathbf{x}_{3}^{-\mathbf{x}} \,\mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x}_{1} - \mathbf{x}_{3}^{-\mathbf{x}} \,\mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x}_{3} \\ & - \,\mathbf{x}_{3}^{-T} \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{B}^{T} \mathbf{x}_{4} - \mathbf{x}_{1}^{-T} \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x}_{1} \\ & - \,\mathbf{x}_{1}^{-T} \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{B}^{T} \mathbf{x}_{2} + \mathbf{x}_{1}^{-T} \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x}_{3}; \end{array}$$

and, finally, for the choice $\mathbf{w}_5 = (-(\mathbf{x_1}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1})^T, \mathbf{0}, (\mathbf{x_3}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1})^T, \mathbf{0})^T$, 123 we obtain 124

$$\mathbf{w_5}^T \mathscr{A} \mathbf{x} = -\mathbf{x_1}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_1} - \mathbf{x_1}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x_3} - \mathbf{x_1}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{B}^T \mathbf{x_4} + \mathbf{x_3}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x_1} + \mathbf{x_3}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{B}^T \mathbf{x_2} - \mathbf{x_3}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_3}.$$

Therefore, we end up with the following expression

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$$\begin{split} (\mathbf{w_1} + \mathbf{w_2} + \mathbf{w_3} + \mathbf{w_4} + \mathbf{w_5})^T \mathscr{A} \mathbf{x} &= \mathbf{x_1}^T \mathbf{M} \mathbf{x_1} + \mathbf{x_3}^T \mathbf{M} \mathbf{x_3} \\ &+ \mathbf{x_1}^T \mathbf{K} \mathbf{x_1} + \mathbf{x_3}^T \mathbf{K} \mathbf{x_3} + \mathbf{x_2}^T \mathbf{A} \mathbf{x_2} + \mathbf{x_4}^T \mathbf{A} \mathbf{x_4} \\ &+ \mathbf{x_4}^T \mathbf{B} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{B}^T \mathbf{x_4} + \mathbf{x_2}^T \mathbf{B} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{B}^T \mathbf{x_2} \\ &- \mathbf{x_3}^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x_3} - \mathbf{x_1}^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x_1} \\ &- \mathbf{x_3}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_3} - \mathbf{x_1}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_1} \\ &- 2\mathbf{x_3}^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_1} + 2\mathbf{x_1}^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_3}. \end{split}$$

For estimating the non-symmetric terms, we use the following result:

$$\begin{aligned} -2\mathbf{x_3}^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_1} &\geq -2 \| (\mathbf{K} + \mathbf{M})^{-1/2} \mathbf{K} \mathbf{x_3} \|_{l_2} \| (\mathbf{K} + \mathbf{M})^{-1/2} \mathbf{M} \mathbf{x_1} \|_{l_2} \\ &\geq - \| (\mathbf{K} + \mathbf{M})^{-1/2} \mathbf{K} \mathbf{x_3} \|_{l_2}^2 - \| (\mathbf{K} + \mathbf{M})^{-1/2} \mathbf{M} \mathbf{x_1} \|_{l_2}^2 \\ &= -\mathbf{x_3}^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x_3} - \mathbf{x_1}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_1}. \end{aligned}$$

Analogously, we obtain

$$2\mathbf{x_1}^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_3} \ge -\mathbf{x_1}^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x_1} - \mathbf{x_3}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_3}$$

Hence, putting all terms together, we have

$$\begin{aligned} (\mathbf{w}_1 + \mathbf{w}_2 + \mathbf{w}_3 + \mathbf{w}_4 + \mathbf{w}_5)^T \mathscr{A} \mathbf{x} &= \mathbf{x}^T \mathscr{C} \mathbf{x} \\ &- 2\mathbf{x}_3^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x}_3 - 2\mathbf{x}_1^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x}_1 \\ &- 2\mathbf{x}_3^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x}_3 - 2\mathbf{x}_1^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x}_1. \end{aligned}$$

In order to get rid of the four remaining terms, we use, for i = 1, 3,

$$\mathbf{x_i}^T \mathbf{K} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{K} \mathbf{x_i} \leq \mathbf{x_i}^T \mathbf{K} \mathbf{x_i} \quad \text{and} \quad \mathbf{x_i}^T \mathbf{M} (\mathbf{K} + \mathbf{M})^{-1} \mathbf{M} \mathbf{x_i} \leq \mathbf{x_i}^T \mathbf{M} \mathbf{x_i}.$$

Hence by adding w_1 and w_2 twice more, we end up with the desired result

$$\underbrace{(\mathbf{3}\mathbf{w}_1 + \mathbf{3}\mathbf{w}_2 + \mathbf{w}_3 + \mathbf{w}_4 + \mathbf{w}_5)^T}_{:=\mathbf{w}^T} \mathscr{A}\mathbf{x} \ge \mathbf{x}^T \mathscr{C}\mathbf{x} + \mathbf{x_2}^T \mathbf{A}\mathbf{x_2} + \mathbf{x_4}^T \mathbf{A}\mathbf{x_4} \ge \mathbf{x}^T \mathscr{C}\mathbf{x}.$$

The next step is to compute (and estimate) the $\mathscr C$ norm of the special test vector. ¹³¹ Straightforward estimations yield ¹³²

$$\|\mathbf{w}\|_{\mathscr{C}}^{2} = \|3\mathbf{w}_{1} + 3\mathbf{w}_{2} + \mathbf{w}_{3} + \mathbf{w}_{4} + \mathbf{w}_{5}\|_{\mathscr{C}}^{2} \le 7 \|\mathbf{x}\|_{\mathscr{C}}^{2}.$$

This completes the proof.

Now, from Lemma 1, we obtain that the condition number of the preconditioned 134 system can be estimated by the constant $c = 2\sqrt{7}$ that is obviously independent of 135 the meshsize *h* and all involved parameters ω , *v* and σ , i.e. 136

$$\kappa_{\mathscr{C}}(\mathscr{C}^{-1}\mathscr{A}) := \|\mathscr{C}^{-1}\mathscr{A}\|_{\mathscr{C}}\|_{\mathscr{C}} \|\mathscr{A}^{-1}\mathscr{C}\|_{\mathscr{C}} \le 2\sqrt{7}.$$
(6)

The condition number defines the convergence behaviour of the MinRes method 137 applied to the preconditioned system (see e.g. [9]), as stated in the following theorem: 138

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Theorem 1 (Robust solver). *The MinRes method applied to the preconditioned sys-* 140 tem $C^{-1} \mathscr{A} \mathbf{u} = C^{-1} \mathbf{f}$ converges. At the 2*m*-th iteration, the preconditioned residual 141 $\mathbf{r}^{\mathbf{m}} = C^{-1} \mathbf{f} - C^{-1} \mathscr{A} \mathbf{u}^{\mathbf{m}}$ is bounded as 142

$$\left\|\mathbf{r}^{2\mathbf{m}}\right\|_{\mathscr{C}} \le \frac{2q^m}{1+q^{2m}} \left\|\mathbf{r}^{\mathbf{0}}\right\|_{\mathscr{C}}, \quad where \quad q = \frac{2\sqrt{7}-1}{2\sqrt{7}+1}.$$
(7)

4 Conclusion, Outlook and Acknowledgments

The method developed in this work shows great potential for solving time-harmonic 144 eddy current problems in an unbounded domain in a robust way. The solution of a 145 fully coupled 4 × 4 block-system can be reduced to the solution of a block-diagonal 146 matrix, where each block corresponds to standard problems. We mention, that by 147 analogous procedure, we can state another robust block-diagonal preconditioner $\tilde{\mathcal{C}} = 148$ diag ($\tilde{\mathcal{J}}_{FEM}, \tilde{\mathcal{J}}_{BEM}, \tilde{\mathcal{J}}_{FEM}, \tilde{\mathcal{J}}_{BEM}$), with $\tilde{\mathcal{J}}_{FEM} = \mathbf{M} + \mathbf{K} + \mathbf{B}^T \tilde{\mathcal{J}}_{BEM}^{-1} \mathbf{B}$ and $\tilde{\mathcal{J}}_{BEM} = 149$ **A**, leading to a condition number bound of 4, see e.g. [15].

Of course this block-diagonal preconditioner is only a theoretical one, since the 151 exact solution of the diagonal blocks corresponding to a standard FEM discretized 152 stationary problem and the Schur-complement of a standard FEM-BEM discretized 153 stationary problem are still prohibitively expensive. Nevertheless, as for the FEM 154 discretized version in [13], this theoretical preconditioner allows us replace the solution of a time-dependent problem by the solution of a sequence of time-independent 156 problems in a robust way, i.e. independent of the space and time discretization pa- 157 rameters h and ω and all additional "bad" parameters. Therefore, the issue of finding 158 robust solvers for the fully coupled time-harmonic system matrix *A* can be reduced 159 to finding robust solvers for the blocks \mathscr{I}_{FEM} and \mathscr{I}_{BEM} , or \mathscr{I}_{FEM} and \mathscr{I}_{BEM} . By 160 replacing these diagonal blocks by standard preconditioners, it is straight-forward 161 to derive mesh-independent convergence rates, see, e.g., [8]. Unfortunately, the con- 162 struction of fully robust preconditioners for the diagonal blocks is not straight for- 163 ward and has to be studied. Candidates are H matrix, multigrid multigrid and do- 164 main decomposition preconditioners, see, e.g. [2, 6] and [12], respectively. 165

The preconditioned MinRes solver presented in this paper can also be generalized 166 to eddy current optimal control problems studied in [14] for the pure FEM case in 167 bounded domains. 168

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